# Logic and Computer Design Fundamentals Chapter 5 - Sequential Circuits 

## Part 1 - Storage Elements and Sequential

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## Overview

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## Introduction to Sequential Circuits

- A Sequential circuit contains:
- Storage elements: Latches or Flip-Flops
- Combinational Logic:
- Implements a multiple-output switching function

- Inputs are signals from the outside.
- Outputs are signals to the outside.
- Other inputs, State or Present State, are signals from storage elements.
- The remaining outputs, Next State are inputs to storage elements.


## Introduction to Sequential Circuits



## Types of Sequential Circuits

- Depends on the times at which:
- storage elements observe their inputs, and
- storage elements change their state
- Synchronous
- Behavior defined from knowledge of its signals at discrete instances of time
- Storage elements observe inputs and can change state only in relation to a timing signal (clock pulses from a clock)
- Asynchronous
- Behavior defined from knowledge of inputs an any instant of time and the order in continuous time in which inputs change
- If clock just regarded as another input, all circuits are asynchronous!
- Nevertheless, the synchronous abstraction makes complex designs tractable!


## Discrete Event Simulation

- In order to understand the time behavior of a sequential circuit we use discrete event simulation.
- Rules:
- Gates modeled by an ideal (instantaneous) function and a fixed gate delay
- Any change in input values is evaluated to see if it causes a change in output value
- Changes in output values are scheduled for the fixed gate delay after the input change
- At the time for a scheduled output change, the output value is changed along with any inputs it drives


## Simulated NAND Gate

- Example: A 2-Input NAND gate with a 0.5 ns. delay:

- Assume A and B have been 1 for a long time
- At time $\mathbf{t}=\mathbf{0}$, A changes to a 0 at $\mathbf{t}=\mathbf{0 . 8}$ ns, back to 1 .

| t (ns) | A | B | F(I) | F | Comment |
| :---: | :---: | :---: | :---: | :---: | :---: |
| - | 1 | 1 | 0 | 0 | $A=B=1$ for a long time |
| 0 | $1 \Rightarrow 0$ | 1 | $1 \Leftarrow 0$ | 0 | F(I) changes to 1 |
| 0.5 | 0 | 1 | 1 | $1 \Leftarrow 0$ | F changes to 1 after a 0.5 ns delay |
| 0.8 | $1 \Leftarrow 0$ | 1 | $1 \Rightarrow 0$ | 1 | F(Instantaneous) changes to 0 |
| 0.13 | 1 | 1 | 0 | $1 \Rightarrow 0$ | $F$ changes to 0 after a 0.5 ns delay |
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## Gate Delay Models

- Suppose gates with delay $n$ ns are represented for $n=0.2 \mathrm{~ns}, \boldsymbol{n}=0.4 \mathrm{~ns}$, $n=0.5 \mathrm{~ns}$, respectively:



## Circuit Delay Model

- Consider a simple A 2-input multiplexer:
- With function:
- $Y=A$ for $S=1$
- $\mathbf{Y}=\mathbf{B}$ for $\mathbf{S}=\mathbf{0}$

- "Glitch" is due to delay of inverter-


## Storing State

- What if A connected to $Y$ ?
- Circuit becomes:
- With function:
- $Y=B$ for $S=1$, and
$Y(t)$ dependent on $Y(t-0.9)$ for $S=0$
B

- The simple combinational circuit has now become a sequential circuit because its output is a function of a time sequence of input signals!


## Storing State (Continued)

- Simulation example as input signals change with time. Changes occur every 100 ns , so that the tenths of ns delays are negligible.
Time

| B | S | Y | Comment |
| :---: | :---: | :---: | :---: |
| 1 | 0 | 0 | Y "remembers" 0 |
| 1 | 1 | 1 | $\mathrm{Y}=\mathrm{B}$ when $\mathrm{S}=1$ |
| 1 | 0 | 1 | Now $Y$ "remembers" $\mathrm{B}=1$ for S = 0 |
| 0 | 0 | 1 | No change in $Y$ when $B$ changes |
| 0 | 1 | 0 | $\mathrm{Y}=\mathrm{B}$ when $\mathrm{S}=1$ |
| 0 | 0 | 0 | Y "remembers" B=0 for S = 0 |
| 1 | 0 | 0 | No change in $Y$ when $B$ changes |

- $Y$ represent the state of the circuit, not just an output.


## Storing State (Continued)

- Suppose we place an inverter in the "feedback path."

- The following behavior results:
- The circuit is said to be unstable.
- For S = 0, the circuit has become what is called an oscillator. Can be

| $\mathbf{B}$ | S | $\mathbf{Y}$ | Comment |
| :---: | :---: | :---: | :--- |
| $\mathbf{0}$ | 1 | 0 | $\mathbf{Y}=\mathbf{B}$ when $\mathrm{S}=1$ |
| 1 | 1 | 1 |  |
| 1 | 0 | 1 | Now $\mathbf{Y}$ "remembers" $\mathbf{A}$ |
| 1 | 0 | 0 | $\mathbf{Y}, 1.1$ ns later |
| 1 | 0 | 1 | $\mathbf{Y}, 1.1$ ns later |
| 1 | 0 | 0 | $\mathbf{Y}, 1.1$ ns later | used as crude clock.

## Basic (NAND) $\overline{\mathbf{S}}-\overline{\mathbf{R}}$ Latch

- "Cross-Coupling" two NAND gates gives the $\overline{\mathrm{S}}-\overline{\mathrm{R}}$ Latch:
- Which has the time
R (reset)
 sequence behavior:
- $S=0, R=0$ is forbidden as

| $\mathbf{R}$ | $\mathbf{S}$ | $\mathbf{Q}$ | $\overline{\mathbf{Q}}$ | Comment |
| :---: | :---: | :---: | :---: | :--- |
| $\mathbf{1}$ | 1 | $?$ | $?$ | Stored state unknown |
| $\mathbf{1}$ | 0 | 1 | 0 | "Set" Q to 1 |
| $\mathbf{1}$ | 1 | 1 | 0 | Now Q "remembers" 1 |
| $\mathbf{0}$ | 1 | 0 | 1 | "Reset" Q to 0 |
| 1 | 1 | 0 | 1 | Now Q "remembers" 0 |
| $\mathbf{0}$ | 0 | 1 | 1 | Both go high |
| 1 | 1 | $?$ | $?$ | Unstable! |

## Basic (NOR) S - R Latch

- Cross-coupling two $\quad$ (reset) NOR gates gives the S - R Latch:
- Which has the time

S (set)
 sequence behavior:


| $R$ | $S$ | $\mathbf{Q}$ | $\overline{\mathbf{Q}}$ | Comment |
| :---: | :---: | :---: | :--- | :--- |
| $\mathbf{0}$ | $\mathbf{0}$ | $?$ | $?$ | Stored state unknown |
| $\mathbf{0}$ | 1 | 1 | $\mathbf{0}$ | "Set" Q to 1 |
| $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{0}$ | Now Q "remembers" $\mathbf{1}$ |
| $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{1}$ | "Reset" Q to 0 |
| $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{1}$ | Now Q "remembers" 0 |
| $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ | Both go low |
| $\mathbf{0}$ | $\mathbf{0}$ | ? | ? | Unstable! |

## Clocked S - R Latch

- Adding two NAND gates to the basic $\bar{S}-\overline{\mathrm{R}}$ NAND latch gives the clocked S - R latch:

- Has a time sequence behavior similar to the basic S-R latch except that the $S$ and $R$ inputs are only observed when the line $C$ is high.
- C means "control" or "clock".


## Clocked S - R Latch (continued)

- The Clocked S-R Latch can be described by a table:

| $Q(t)$ | $S$ | $R$ | $Q(t+1)$ | Comment |
| :---: | :---: | :---: | :---: | :--- |
|  | 0 | 0 | 0 | 0 | No change based on:

- current inputs (S,R) and
- current state $\mathbf{Q}(\mathbf{t})$.


## D Latch

- Adding an inverter


The graphic symbol for a D Latch is:


## Flip-Flops

- The latch timing problem
- Master-slave flip-flop
- Edge-triggered flip-flop
- Standard symbols for storage elements
- Direct inputs to flip-flops


## The Latch Timing Problem

- In a sequential circuit, paths may exist through combinational logic:
- From one storage element to another
- From a storage element back to the same storage element
- The combinational logic between a latch output and a latch input may be as simple as an interconnect
- For a clocked D-latch, the output $\mathbf{Q}$ depends on the input $D$ whenever the clock input $C$ has value 1


## The Latch Timing Problem (continued)

- Consider the following circuit:
- Suppose that initially $\mathbf{Y}=\mathbf{0}$.


Clock $\qquad$
Y $\qquad$

- As long as $\mathbf{C = 1}$, the value of $Y$ continues to change!
- The changes are based on the delay present on the loop through the connection from $Y$ back to $Y$.
- This behavior is clearly unacceptable.
- Desired behavior: Y changes only once per clock pulse


## The Latch Timing Problem (continued)

- A solution to the latch timing problem is to break the closed path from $Y$ to $Y$ within the storage element
- The commonly-used, path-breaking solutions replace the clocked D-latch with:
- a master-slave flip-flop
- an edge-triggered flip-flop


## S-R Master-Slave Flip-Flop

- Consists of two clocked S-R latches in series with the clock on the second latch inverted
- The input is observed by the first latch with $\mathrm{C}=1$

- The output is changed by the second latch with $C=0$
- The path from input to output is broken by the difference in clocking values ( $\mathrm{C}=1$ and $\mathrm{C}=0$ ).
- The behavior demonstrated by the example with $D$ driven by $Y$ given previously is prevented since the clock must change from 1 to 0 before a change in $Y$ based on D can occur.


## Flip-Flop Problem

- The change in the flip-flop output is delayed by the pulse width which makes the circuit slower or
- $S$ and/or $R$ are permitted to change while $C=1$
- Suppose Q = 0 and $S$ goes to 1 and then back to 0 with $R$ remaining at 0
- The master latch sets to 1
- A 1 is transferred to the slave
- Suppose $\mathbf{Q}=0$ and $S$ goes to 1 and back to 0 and $R$ goes to 1 and back to 0
- The master latch sets and then resets
- A 0 is transferred to the slave
- This behavior is called $1 s$ catching


## Flip-Flop Solution

- Use edge-triggering instead of master-slave
- An edge-triggered flip-flop ignores the pulse while it is at a constant level and triggers only during a transition of the clock signal
- Edge-triggered flip-flops can be built directly at the electronic circuit level, or
- A master-slave D flip-flop which also exhibits edge-triggered behavior can be used.


## Edge-Triggered D Flip-Flop

- The edge-triggered D flip-flop is the same as the masterslave $D$ flip-flop
- It can be formed by:
- Replacing the first clocked S-R latch with a clocked D latch or
- Adding a D input and inverter to a master-slave S-R flip-flop
- The delay of the S-R master-slave flip-flop can be avoided since the 1 s -catching behavior is not present with $D$ replacing $S$ and $R$ inputs
- The change of the $\mathbf{D}$ flip-flop output is associated with the negative edge at the end of the pulse
- It is called a negative-edge triggered flip-flop


## Positive-Edge Triggered D Flip-Flop

- Formed by adding inverter to clock input

- $Q$ changes to the value on $D$ applied at the positive clock edge within timing constraints to be specified
- Our choice as the standard flip-flop for most sequential circuits


## Standard Symbols for Storage Elements



- Master-Slave:
(a) Latches



## Direct Inputs

- At power up or at reset, all or part of a sequential circuit usually is initialized to a known state before it begins operation
- This initialization is often done outside of the clocked behavior of the circuit, i.e., asynchronously.

- Direct $R$ and/or $S$ inputs that control the state of the latches within the flip-flops are used for this initialization.
- For the example flip-flop shown
- 0 applied to $\overline{\mathbf{R}}$ resets the flip-flop to the $\mathbf{0}$ state
- 0 applied to $\overline{\mathbf{S}}$ sets the flip-flop to the $\mathbf{1}$ state


## Sequential Circuit Analysis

- General Model
- Current State at time ( $t$ ) is stored in an array of flip-flops.
- Next State at time (t+1) is a Boolean function of CLK $\longrightarrow$ State and Inputs.
- Outputs at time ( $\mathbf{t}$ ) are a Boolean function of State (t) and (sometimes) Inputs (t).


## Example 1 (from Fig. 5-15)

- Input: $\quad \mathbf{x}(\mathbf{t})$
- Output: $\quad \mathbf{y}(\mathbf{t})$
- State: (A(t), B(t))
- What is the Output Function?
- What is the Next State
 Function?


## Example 1 (from Fig. 5-15) (continued)

- Boolean equations for the functions:
- $\mathbf{A}(\mathbf{t}+\mathbf{1})=\mathbf{A}(\mathbf{t}) \mathbf{x}(\mathbf{t})$ $+\mathbf{B}(\mathbf{t}) \mathbf{x}(\mathbf{t})$
- $\mathbf{B}(\mathbf{t}+\mathbf{1})=\overline{\mathbf{A}}(\mathbf{t}) \mathbf{x}(\mathbf{t})$
- $\mathbf{y}(\mathrm{t})=\overline{\mathbf{x}}(\mathrm{t})(\mathbf{B}(\mathrm{t})+\mathbf{A}(\mathrm{t}))$



## Example 1(from Fig. 5-15) (continued)

- Where in time are inputs, outputs and states defined?



## State Table Characteristics

- State table - a multiple variable table with the following four sections:
- Present State - the values of the state variables for each allowed state.
- Input - the input combinations allowed.
- Next-state - the value of the state at time ( $\mathrm{t}+1$ ) based on the present state and the input.
- Output - the value of the output as a function of the present state and (sometimes) the input.
- From the viewpoint of a truth table:
- the inputs are Input, Present State
- and the outputs are Output, Next State


## Example 1: State Table (from Fig. 5-15)

- The state table can be filled in using the next state and output equations: $\mathbf{A}(\mathbf{t}+1)=\mathbf{A}(\mathbf{t}) \mathbf{x}(\mathbf{t})+\mathbf{B}(\mathbf{t}) \mathbf{x}(\mathbf{t}) \mathbf{B}(\mathbf{t}+1)$
$=\overline{\mathbf{A}}(\mathbf{t}) \mathbf{x}(\mathbf{t})$
$\mathbf{y}(\mathrm{t})$
$=\mathbf{x}(\mathbf{t})(\mathbf{B}(\mathbf{t})+\mathbf{A}(\mathbf{t}))$

| Present State | Input | Next State |  | Output |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{A}(\mathbf{t})$ | $\mathbf{B}(\mathbf{t})$ | $\mathbf{x}(\mathbf{t})$ | $\mathbf{A}(\mathbf{t}+\mathbf{1})$ | $\mathbf{B}(\mathbf{t}+\mathbf{1})$ |
| $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{y}(\mathbf{t})$ |  |
| $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{1}$ | $\mathbf{0}$ |

## Example 1: Alternate State Table

- 2-dimensional table that matches well to a K-map. Present state rows and input columns in Gray code order.
- $\mathbf{A}(\mathbf{t}+\mathbf{1})=\mathbf{A}(\mathbf{t}) \mathbf{x}(\mathbf{t})+\mathbf{B}(\mathbf{t}) \mathbf{x}(\mathbf{t})$
- $\mathbf{B}(\mathbf{t}+\mathbf{1})=\overline{\mathbf{A}}(\mathbf{t}) \mathbf{x}(\mathbf{t})$
- $\mathbf{y}(\mathrm{t})=\overline{\mathrm{x}}(\mathrm{t})(\mathrm{B}(\mathrm{t})+\mathrm{A}(\mathrm{t}))$

| Present State $A(t) B(t)$ | Next State |  | Output |  |
| :---: | :---: | :---: | :---: | :---: |
|  | $\mathrm{x}(\mathrm{t})=0$ | $\mathrm{x}(\mathrm{t})=1$ | $x(t)=0$ | $x(t)=1$ |
|  | $\mathbf{A}(\mathrm{t}+1) \mathrm{B}(\mathrm{t}+1)$ | $\mathbf{A}(\mathbf{t}+1) \mathbf{B}(\mathbf{t}+1)$ | y(t) | $\mathrm{y}(\mathrm{t})$ |
| 00 | 00 | 01 | 0 | 0 |
| 01 | 00 | 11 | 1 | 0 |
| 10 | 00 | 10 | 1 | 0 |
| 11 | 00 | 10 | 1 | 0 |

## State Diagrams

- The sequential circuit function can be represented in graphical form as a state diagram with the following components:
- A circle with the state name in it for each state
- A directed arc from the Present State to the Next State for each state transition
- A label on each directed arc with the Input values which causes the state transition, and
- A label:
- On each circle with the output value produced, or
- On each directed arc with the output value produced.


## State Diagrams

- Label form:
- On circle with output included:
- state/output
- Moore type output depends only on state
- On directed arc with the output included:
- input/output
- Mealy type output depends on state and input


## Example 1: State Diagram

- Which type?
- Diagram gets confusing for large circuits
- For small circuits, usually easier to understand than the state table



## Equivalent State Definitions

- Two states are equivalent if their response for each possible input sequence is an identical output sequence.
- Alternatively, two states are equivalent if their outputs produced for each input symbol is identical and their next states for each input symbol are the same or equivalent.


## Equivalent State Example

- Text Figure 5-17(a):
- For states S3 and S2,
- the output for input 0 is 1 and input 1 is $0,0 / 1$ and
- the next state for input 0 is SO and for input 1 is S 2 .

- By the alternative definition, states S3 and S2 are equivalent.


## Equivalent State Example

- Replacing S3 and S2 by a single state gives state diagram:
- Examining the new diagram, states S1 and S2 are equivalent since
- their outputs for input 0 is 1 and input 1 is 0 , and
- their next state for input 0 is S 0 and for input 1 is $S 2$,
- Replacing S1 and S2 by a single state gives state diagram:



## Moore and Mealy Models

- Sequential Circuits or Sequential Machines are also called Finite State Machines (FSMs). Two formal models exist:
- Moore Model
- Mealy Model
- Named after E.F. Moore - Named after G. Mealy
- Outputs are a function ONLY of states
- Usually specified on the states.
- Outputs are a function of inputs AND states
- Usually specified on the state transition arcs.


## Moore and Mealy Example Diagrams

- Mealy Model State Diagram maps inputs and state to outputs

- Moore Model State Diagram maps states to outputs



## Moore and Mealy Example Tables

- Moore Model state table maps state to outputs

| Present | Next State |  | Output |
| :---: | :---: | :---: | :---: |
| State | $\mathbf{x}=\mathbf{0}$ | $\mathbf{x}=1$ |  |
| 0 | 0 | 1 | 0 |
| 1 | 0 | 2 | 0 |
| 2 | 0 | 2 | 1 |

- Mealy Model state table maps inputs and state to outputs $\operatorname{Present}$ Next State Output

| State | $\mathrm{x}=\mathbf{0}$ | $\mathrm{x}=\mathbf{1}$ | $\mathrm{x}=\mathbf{0}$ | $\mathrm{x}=\mathbf{1}$ |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{0}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{0}$ |
| $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{1}$ | $\mathbf{0}$ | $\mathbf{1}$ |

## Mixed Moore and Mealy Outputs

- In real designs, some outputs may be Moore type and other outputs may be Mealy type.
- Example: Figure 5-17(a) can be modified to illustrate this
- State 00: Moore
- States 01, 10, and 11: Mealy
- Simplifies output specification



## Example 2: Sequential Circuit Analysis



## Example 2: Flip-Flop Input Equations

- Variables
- Inputs: None
- Outputs: Z
- State Variables: A, B, C
- Initialization: Reset to (0,0,0)
- Equations
- $\mathbf{A}(\mathbf{t}+\mathbf{1})=\quad \mathbf{Z}=$
- $\mathbf{B}(\mathbf{t}+\mathbf{1})=$
- $\mathbf{C}(\mathbf{t}+\mathbf{1})=$


## Example 2: State Table

$$
\mathbf{X}^{\prime}=\mathbf{X}(\mathbf{t}+\mathbf{1}) \begin{array}{|ccc|c|c|}
\hline \text { A B C C } & \text { A' }^{\prime} \mathbf{B}^{\prime} C^{\prime} & \mathbf{Z} \\
\hline 0 & 0 & 0 & & \\
\hline 0 & 0 & 1 & & \\
\hline 0 & 1 & 0 & & \\
\hline 0 & 1 & 1 & & \\
\hline 1 & 0 & 0 & & \\
\hline 1 & 0 & 1 & & \\
\hline 1 & 1 & 0 & & \\
\hline 1 & 1 & 1 & & \\
\hline
\end{array}
$$

## Example 2: State Diagram



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