# Logic and Computer Design Fundamentals <br> Chapter 3 - Combinational Logic Design 

Part 2 - Combinational Logic

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## Overview

- Part 2 - Combinational Logic
- Functions and functional blocks
- Rudimentary logic functions
- Decoding using Decoders
- Implementing Combinational Functions with Decoders
- Encoding using Encoders
- Selecting using Multiplexers
- Implementing Combinational Functions with Multiplexers


## Functions and Functional Blocks

- The functions considered are those found to be very useful in design
- Corresponding to each of the functions is a combinational circuit implementation called a functional block.
- In the past, functional blocks were packaged as small-scale-integrated (SSI), medium-scale integrated (MSI), and large-scale-integrated (LSI) circuits.
- Today, they are often simply implemented within a very-large-scale-integrated (VLSI) circuit.


## Rudimentary Logic Functions

- Functions of a single variable $X$
- Can be used on the inputs to functional blocks to implement other than the block's intended function

TABLE 4-1
Functions of One Variable

| $\mathbf{X}$ | $\mathbf{F = 0}$ | $\mathbf{F}=\mathbf{X}$ | $\mathbf{F}=\overline{\mathbf{X}}$ | $\mathbf{F = 1}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 | 1 |
| 1 | 0 | 1 | 0 | 1 |

(c)


(b)

(d)

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(a)


## Multiple-bit Rudimentary Functions


(a)
(b)

- A wide line is used to represent a bus which is a vector signal


(d)
- In (b) of the example, $F=\left(F_{3}, F_{2}, F_{1}, F_{0}\right)$ is a bus.
- The bus can be split into individual bits as shown in (b)
- Sets of bits can be split from the bus as shown in (c) for bits 2 and 1 of $F$.
- The sets of bits need not be continuous as shown in (d) for bits 3 , 1 , and 0 of $\mathbf{F}$.


## Enabling Function

- Enabling permits an input signal to pass through to an output
- Disabling blocks an input signal from passing through to an output, replacing it with a fixed value
- The value on the output when it is disable can be Hi-Z (as for three-state buffers and transmission gates), 0 , or $1 \underset{\mathrm{EN} \longrightarrow \mathrm{F}}{\mathrm{X} \longrightarrow}$
- When disabled, 0 output
(a)
- When disabled, 1 output
- See Enabling App in text

(b)

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## Decoding

- Decoding - the conversion of an n-bit input code to an $m$-bit output code with $\mathbf{n} \leq \mathbf{m} \leq \mathbf{2}^{\boldsymbol{n}}$ such that each valid code word produces a unique output code
- Circuits that perform decoding are called decoders
- Here, functional blocks for decoding are
- called $n$-to- $m$ line decoders, where $m \leq 2^{n}$, and
- generate $2^{\boldsymbol{n}}$ (or fewer) minterms for the $\boldsymbol{n}$ input variables


## Decoder Examples

- 1-to-2-Line Decoder $A \mid D_{0} \mathbf{D}_{\mathbf{1}}$

- 2-to-4-Line Decoder

(a)
- Note that the 2-4-line made up of 21-to-2line decoders and 4 AND gates.


## Decoder Expansion

- General procedure given in book for any decoder with $n$ inputs and $2^{n}$ outputs.
- This procedure builds a decoder backward from the outputs.
- The output AND gates are driven by two decoders with their numbers of inputs either equal or differing by 1.
- These decoders are then designed using the same procedure until 2-to-1-line decoders are reached.
- The procedure can be modified to apply to decoders with the number of outputs $\neq \mathbf{2}^{\text {n }}$


## Decoder Expansion - Example 1

- 3-to-8-line decoder
- Number of output ANDs $=8$
- Number of inputs to decoders driving output ANDs = 3
- Closest possible split to equal
- 2-to-4-line decoder
- 1-to-2-line decoder
- 2-to-4-line decoder
- Number of output ANDs = 4
- Number of inputs to decoders driving output ANDs = 2
- Closest possible split to equal
- Two 1-to-2-line decoders
- See next slide for result


## Decoder Expansion - Example 1

- Result



## Decoder Expansion - Example 2

- 7-to-128-line decoder
- Number of output ANDs = 128
- Number of inputs to decoders driving output ANDs $=7$
- Closest possible split to equal
- 4-to-16-line decoder
- 3-to-8-line decoder
- 4-to-16-line decoder
- Number of output ANDs = 16
- Number of inputs to decoders driving output ANDs = 2
- Closest possible split to equal
- 2 2-to-4-line decoders
- Complete using known 3-8 and 2-to-4 line decoders


## Decoder with Enable

- In general, attach m-enabling circuits to the outputs
- See truth table below for function
- Note use of X's to denote both 0 and 1
- Combination containing two X's represent four binary combinations
- Alternatively, can be viewed as distributing value of signal EN to 1 of 4 outputs
- In this case, called a demultiplexer

(a)

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(b)

Logic and Computer Design Fundamentals, 4e
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## Combinational Logic Implementation - Decoder and OR Gates

- Implement $\boldsymbol{m}$ functions of $\boldsymbol{n}$ variables with:
- Sum-of-minterms expressions
- One n-to-2 ${ }^{n}$-line decoder
- m OR gates, one for each output
- Approach 1:
- Find the truth table for the functions
- Make a connection to the corresponding OR from the corresponding decoder output wherever a 1 appears in the truth table
- Approach 2
- Find the minterms for each output function


## Decoder and OR Gates Example

- Implement the following set of odd parity functions of $\left(\mathbf{A}_{7}, \mathbf{A}_{6}, \mathbf{A}_{5}, \mathbf{A}_{3}\right)$ $\mathbf{P}_{1}=\mathbf{A}_{7} \oplus \mathbf{A}_{5} \oplus \mathbf{A}_{3}$ $\mathbf{P}_{2}=\mathbf{A}_{7} \oplus \mathbf{A}_{6} \oplus \mathbf{A}_{3}$ $\mathbf{P}_{4}=\mathbf{A}_{7} \oplus \mathbf{A}_{6} \oplus \mathbf{A}_{5}$
- Finding sum of minterms expressions
$P_{1}=\Sigma_{\mathrm{m}}(1,2,5,6,8,11,12,15)$
$P_{2}=\Sigma_{m}(1,3,4,6,8,10,13,15)$
$P_{4}=\Sigma_{m}(2,3,4,5,8,9,14,15)$
- Find circuit
- Is this a good idea?



## Encoding

- Encoding - the opposite of decoding - the conversion of an $m$-bit input code to a $n$-bit output code with $n \leq$ $\boldsymbol{m} \leq 2^{\boldsymbol{n}}$ such that each valid code word produces a unique output code
- Circuits that perform encoding are called encoders
- An encoder has $2^{\boldsymbol{n}}$ (or fewer) input lines and $n$ output lines which generate the binary code corresponding to the input values
- Typically, an encoder converts a code containing exactly one bit that is 1 to a binary code corresponding to the position in which the 1 appears.


## Encoder Example

## - A decimal-to-BCD encoder

- Inputs: 10 bits corresponding to decimal digits 0 through $9,\left(D_{0}, \ldots, D_{9}\right)$
- Outputs: 4 bits with BCD codes
- Function: If input bit $D_{i}$ is a 1 , then the output ( $A_{3}, A_{2}, A_{1}, A_{0}$ ) is the BCD code for $i$,
- The truth table could be formed, but alternatively, the equations for each of the four outputs can be obtained directly.


## Encoder Example (continued)

- Input $D_{i}$ is a term in equation $A_{j}$ if bit $A_{j}$ is 1 in the binary value for $i$.
- Equations:
$\mathrm{A}_{3}=\mathrm{D}_{8}+\mathrm{D}_{9}$
$\mathrm{A}_{2}=\mathrm{D}_{4}+\mathrm{D}_{5}+\mathrm{D}_{6}+\mathrm{D}_{7}$
$\mathrm{A}_{1}=\mathrm{D}_{2}+\mathrm{D}_{3}+\mathrm{D}_{6}+\mathrm{D}_{7}$
$\mathrm{A}_{0}=\mathrm{D}_{1}+\mathrm{D}_{3}+\mathrm{D}_{5}+\mathrm{D}_{7}+\mathrm{D}_{9}$
- $F_{1}=D_{6}+D_{7}$ can be extracted from $A_{2}$ and $A_{1}$ Is there any cost saving?


## Priority Encoder

- If more than one input value is 1 , then the encoder just designed does not work.
- One encoder that can accept all possible combinations of input values and produce a meaningful result is a priority encoder.
- Among the 1s that appear, it selects the most significant input position (or the least significant input position) containing a 1 and responds with the corresponding binary code for that position.


## Priority Encoder Example

- Priority encoder with 5 inputs $\left(D_{4}, D_{3}, D_{2}, D_{1}, D_{0}\right)$ - highest priority to most significant 1 present - Code outputs A2, A1, A0 and $V$ where $V$ indicates at least one 1 present.

| No. of Min- <br> terms/Row | Inputs |  |  |  |  | Outputs |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D4 | D3 | D2 | D1 | D0 | A2 | A1 | A0 | V |
| 1 | 0 | 0 | 0 | 0 | 0 | X | X | X | 0 |
| 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 |
| 2 | 0 | 0 | 0 | 1 | $X$ | 0 | 0 | 1 | 1 |
| 4 | 0 | 0 | 1 | $X$ | $X$ | 0 | 1 | 0 | 1 |
| 8 | 0 | 1 | $X$ | $X$ | $X$ | 0 | 1 | 1 | 1 |
| 16 | 1 | $X$ | $X$ | $X$ | $X$ | 1 | 0 | 0 | 1 |

- Xs in input part of table represent 0 or 1 ; thus table entries correspond to product terms instead of minterms. The column on the left shows that all 32 minterms are present in the product terms in the table


## Priority Encoder Example (continued)

- Could use a K-map to get equations, but can be read directly from table and manually optimized if careful:

$$
\begin{aligned}
& A_{2}=\mathbf{D}_{4} \\
& A_{1}=\overline{\mathbf{D}}_{4} \mathbf{D}_{3}+\overline{\mathbf{D}}_{4} \overline{\mathbf{D}}_{3} \mathbf{D}_{2}=\overline{\mathrm{D}}_{4} \mathbf{F}_{1}, \mathrm{~F}_{1}=\left(\mathbf{D}_{3}+\mathbf{D}_{2}\right) \\
& \mathbf{A}_{0}=\overline{\mathbf{D}}_{4} \mathbf{D}_{3}+\overline{\mathbf{D}}_{4} \overline{\mathbf{D}}_{3} \overline{\mathbf{D}}_{2} \mathbf{D}_{1}=\overline{\mathbf{D}}_{4}\left(\mathbf{D}_{3}+\overline{\mathbf{D}}_{2} \mathbf{D} 1\right) \\
& \mathbf{V}=\mathbf{D}_{4}+\mathrm{F}_{1}+\mathrm{D}_{1}+\mathbf{D}_{0}
\end{aligned}
$$

## Selecting

- Selecting of data or information is a critical function in digital systems and computers
- Circuits that perform selecting have:
- A set of information inputs from which the selection is made
- A single output
- A set of control lines for making the selection
- Logic circuits that perform selecting are called multiplexers
- Selecting can also be done by three-state logic or transmission gates


## Multiplexers

- A multiplexer selects information from an input line and directs the information to an output line
- A typical multiplexer has $n$ control inputs $\left(S_{n-1}, \ldots S_{0}\right)$ called selection inputs, $2^{n}$ information inputs ( $\mathrm{I}_{2}{ }^{\mathrm{n}}, \ldots \mathrm{I}_{0}$ ), and one output Y
- A multiplexer can be designed to have $m$ information inputs with $\mathrm{m}<2^{\mathrm{n}}$ as well as $n$ selection inputs


## 2-to-1-Line Multiplexer

- Since 2 = $\mathbf{2}^{1}$, $\mathrm{n}=1$
- The single selection variable $S$ has two values:
- $S=0$ selects input $I_{0}$
- $S=1$ selects input $I_{1}$
- The equation:
$\mathbf{Y}=\overline{\mathbf{S}} \mathbf{I}_{\mathbf{0}}+\mathbf{S I}_{\mathbf{1}}$
- The circuit:


## 2-to-1-Line Multiplexer (continued)

- Note the regions of the multiplexer circuit shown:
- 1-to-2-line Decoder
- 2 Enabling circuits
- 2-input OR gate
- To obtain a basis for multiplexer expansion, we combine the Enabling circuits and OR gate into a $2 \times 2$ AND-OR circuit:
- 1-to-2-line decoder
- $2 \times 2$ AND-OR
- In general, for an $\mathbf{2}^{\boldsymbol{n}}$-to-1-line multiplexer:
- n-to-2 ${ }^{n}$-line decoder
- $2^{n} \times 2$ AND-OR


## Example: 4-to-1-line Multiplexer

- 2-to-2 ${ }^{2}$-line decoder
- $2^{2} \times 2$ AND-OR



## Multiplexer Width Expansion

- Select "vectors of bits" instead of "bits"
- Use multiple copies of $2^{n} \times 2$ AND-OR in parallel
- Example: 4-to-1-line quad multiplexer



## Other Selection Implementations

- Three-state logic in place of AND-OR

(b)
- Gate input cost = 14 compared to 22 (or

18) for gate implementation

## Combinational Logic Implementation <br> - Multiplexer Approach 1

- Implement $\boldsymbol{m}$ functions of $\boldsymbol{n}$ variables with:
- Sum-of-minterms expressions
- An m-wide $\mathbf{2}^{\boldsymbol{n}}$-to-1-line multiplexer
- Design:
- Find the truth table for the functions.
- In the order they appear in the truth table:
- Apply the function input variables to the multiplexer inputs $S_{n-1}, \ldots, S_{0}$
- Label the outputs of the multiplexer with the output variables
- Value-fix the information inputs to the multiplexer using the values from the truth table (for don't cares, apply either 0 or 1 )


## Example: Gray to Binary Code

- Design a circuit to convert a 3-bit Gray code to a binary code
- The formulation gives the truth table on the right
- It is obvious from this

| Gray | Binary |
| :---: | :---: |
| $\mathbf{A B C C}$ | $\mathbf{x y z}$ |
| 000 | 000 |
| 100 | 001 |
| 110 | 010 |
| 010 | 011 |
| 011 | 100 |
| 111 | 101 |
| 101 | 110 |
| 001 | 111 |

table that $X=C$ and the
$Y$ and $Z$ are more complex

## Gray to Binary (continued)

- Rearrange the table so that the input combinations are in counting order
- Functions y and z can be implemented using a dual 8-to-1-line multiplexer by:

| Gray <br> $\mathbf{A B C}$ | Binary <br> $\mathbf{x y ~ z}$ |
| :---: | :---: |
| 000 | 000 |
| 001 | 111 |
| 010 | 011 |
| 011 | 100 |
| 100 | 001 |
| 101 | 110 |
| 110 | 010 |
| 111 | 101 |

- connecting A, B, and $C$ to the multiplexer select inputs
- placing $y$ and $z$ on the two multiplexer outputs
- connecting their respective truth table values to the inputs


## Gray to Binary (continued)



- Note that the multiplexer with fixed inputs is identical to a ROM with 3-bit addresses and 2-bit data!


## Combinational Logic Implementation <br> - Multiplexer Approach 2

- Implement any $\boldsymbol{m}$ functions of $\boldsymbol{n}+1$ variables by using:
- An m-wide $\mathbf{2 n}^{\boldsymbol{n}}$-to-1-line multiplexer
- A single inverter
- Design:
- Find the truth table for the functions.
- Based on the values of the first $\boldsymbol{n}$ variables, separate the truth table rows into pairs
- For each pair and output, define a rudimentary function of the final variable $(0,1, X, \overline{\mathbf{X}})$
- Using the first $n$ variables as the index, value-fix the information inputs to the multiplexer with the corresponding rudimentary functions
- Use the inverter to generate the rudimentary function $\overline{\mathbf{X}}$


## Example: Gray to Binary Code

- Design a circuit to convert a 3-bit Gray code to a binary code
- The formulation gives the truth table on the right
- It is obvious from this

| Gray | Binary |
| :---: | :---: |
| $\mathbf{A B C C}$ | $\mathbf{x y z}$ |
| 000 | 000 |
| 100 | 001 |
| 110 | 010 |
| 010 | 011 |
| 011 | 100 |
| 111 | 101 |
| 101 | 110 |
| 001 | 111 |

table that $X=C$ and the
$Y$ and $Z$ are more complex

## Gray to Binary (continued)

- Rearrange the table so that the input combinations are in counting order, pair rows, and find rudimentary functions

| $\begin{gathered} \text { Gray } \\ \text { A B C } \end{gathered}$ | Binary xyz | Rudimentary <br> Functions of C for $y$ | Rudimentary Functions of C for z |
| :---: | :---: | :---: | :---: |
| 000 | 000 | $\mathbf{F}=\mathbf{C}$ | F $=\mathbf{C}$ |
| 001 | 111 |  |  |
| 010 | 011 | $\mathbf{F}=\overline{\mathbf{C}}$ | $\mathbf{F}=\overline{\mathbf{C}}$ |
| 011 | 100 |  |  |
| 100 | 001 | $\mathrm{F}=\mathbf{C}$ | $\mathbf{F}=\overline{\mathbf{C}}$ |
| 101 | 110 |  |  |
| 110 | 010 | $\mathbf{F}=\overline{\mathbf{C}}$ | $\mathrm{F}=\mathbf{C}$ |
| 111 | 101 |  |  |

## Gray to Binary (continued)

- Assign the variables and functions to the multiplexer inputs:

- Note that this approach (Approach 2) reduces the cost by almost half compared to Approach 1.
- This result is no longer ROM-like
- Extending, a function of more than $n$ variables is decomposed into several sub-functions defined on a subset of the variables. The multiplexer then selects among these sub-functions.


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